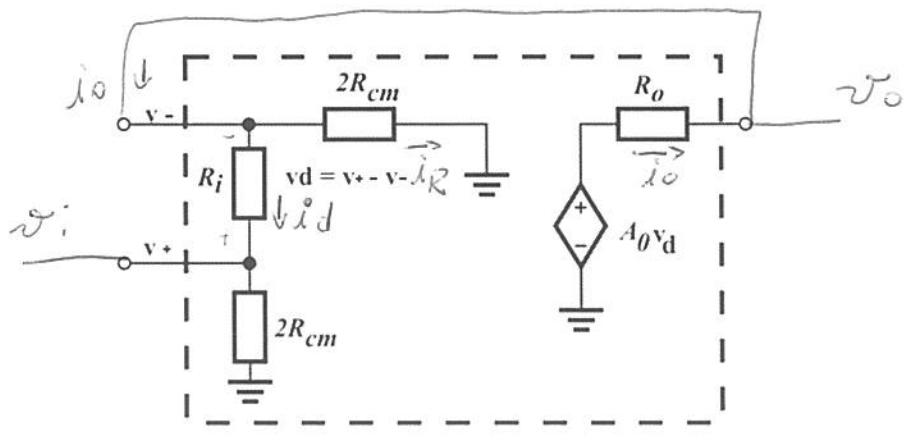


Nome:

RA:

obs: Prova sem consulta. Pode usar calculadora. Respostas a lápis serão corrigidas, mas não aceito reclamação da correção das mesmas.

- 1) a) Deduza a fórmula do ganho de um buffer de tensão. Utilize o modelo do AMP. OP. não ideal da figura abaixo.  
 b) Dados  $R_i = 2M\Omega$ ;  $2R_{CM} = 400M\Omega$ ;  $R_o = 75\Omega$ ;  $A_0 = 10^5$ , calcule o ganho de um buffer construído com o AMP. OP. não ideal.



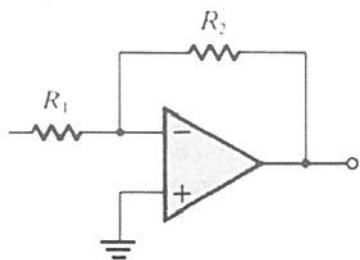
a)

$$\begin{aligned}
 & v_+ = v_i \\
 & v_- = v_o \\
 & v_o = A_0 v_d - R_o i_o \quad \xrightarrow{i_d + i_R} \\
 & v_o = A_0 (v_+ - v_-) - R_o (i_d + i_R) \quad \xrightarrow{v_+ - v_-} \\
 & v_o = A_0 (v_+ - v_-) - R_o (i_d + i_R) = A_0 (v_+ - v_-) - A_0 (v_+ - v_-) R_o / R_i \\
 & v_o = A_0 v_i - A_0 v_o + \frac{R_o v_i}{R_i} - \frac{R_o v_o}{R_i} - \frac{R_o v_o}{2R_{CM}} \\
 & v_o (1 + A_0 + \frac{R_o}{R_i} + \frac{R_o}{2R_{CM}}) = v_i (A_0 + \frac{R_o}{R_i})
 \end{aligned}$$

$$\Rightarrow \frac{v_o}{v_i} = \frac{A_0 + \frac{R_o}{R_i}}{1 + A_0 + \frac{R_o}{R_i} + \frac{R_o}{2R_{CM}}}$$

$$\begin{aligned}
 b) A_v &= \frac{10^5 + \frac{75}{2 \cdot 10^6}}{1 + 10^5 + \frac{75}{2 \cdot 10^6} + \frac{75}{400 \cdot 10^6}} \Rightarrow A_v = 9,9999 \cdot 10^{-1}
 \end{aligned}$$

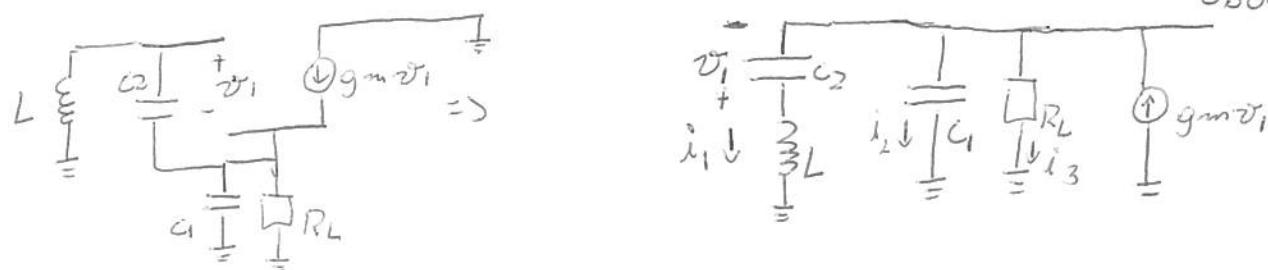
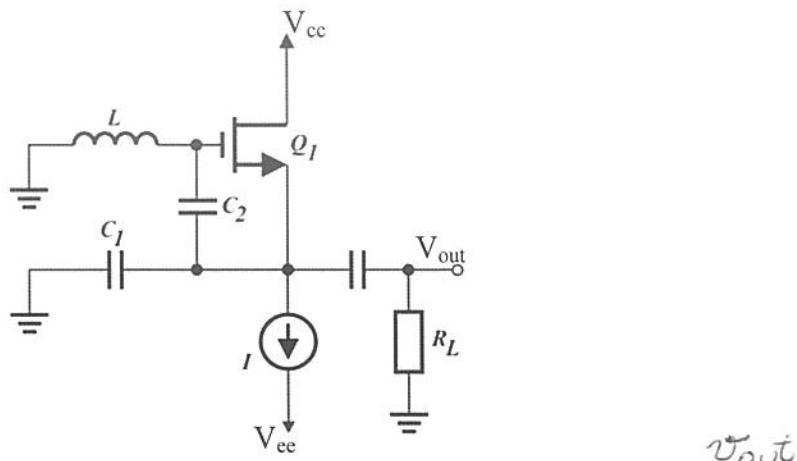
2) Sabendo que  $f_0=1\text{KHZ}$ , determine a frequência de corte aproximada de um amplificador inversor. Dados:  $A_0=10^7$ ;  $R_2=100\text{k}\Omega$ ;  $R_1=10\text{k}\Omega$ .



$$Av = -\frac{R_2}{R_1} = \frac{-100}{10} \Rightarrow Av = -10$$

- Como a frequência é aproximada, podemos utilizar:  $|Av \cdot f_c| = \text{constante}$   
 $|A_0 \cdot f_0| = |Av \cdot f_c|$   
 $\therefore 10^7 \cdot 10^3 = 10 \cdot f_c \Rightarrow \boxed{f_c = 10^9 \text{ Hz}}$

3) Considere  $\lambda=0$  e deduza a fórmula da frequência de oscilação e o valor de  $g_m R$ , de modo que ocorra oscilação.



$$\circ i_1 = \frac{V_{out}(s)}{sL + \frac{1}{sC_2}} = \frac{V_{out}(s)}{\frac{s^2LC_2+1}{sC_2}} \Rightarrow i_1 = \frac{V_{out}(s) \cdot sC_2}{s^2LC_2+1}$$

$$\circ i_2 = \frac{V_{out}(s)}{\frac{1}{sC_1}} \Rightarrow i_2 = sC_1 \frac{V_{out}(s)}{1}$$

$$\circ i_3 = \frac{V_{out}}{R_L} \Rightarrow B(s)$$

$$\circ v_1 = -(sC_1)^{-1} i_1 = -sC_1^{-1} \frac{V_{out}(s) \cdot sC_2}{s^2LC_2+1} \Rightarrow \frac{v_1}{V_{out}(s)} = \frac{-1}{s^2LC_2+1}$$

$$\circ g_m v_1 = i_1 + i_2 + i_3 \Rightarrow g_m v_1 = V_{out}(s) \cdot \frac{sC_2}{s^2LC_2+1} + sC_1 \frac{V_{out}(s)}{R_L} + \frac{V_{out}}{R_L} \Rightarrow \frac{V_{out}}{V_1} = \frac{g_m}{s^2LC_2+1} + \frac{1}{sC_1 + \frac{1}{R_L}}$$

$$\Rightarrow \frac{V_{out}}{V_1} = \frac{g_m}{sRLC_2 + sC_1RL(s^2LC_2+1) + (s^2LC_2+1)} \Rightarrow \frac{V_{out}(s)}{V_1} = \frac{g_m RL(s^2LC_2+1)}{s^3LC_1C_2R_L + s^2LC_2 + s(RLC_2 + CR)}$$

$$\Rightarrow \frac{V_{out}(s)}{V_1} = \frac{g_m RL(s^2LC_2+1)}{s^3LC_1C_2R_L + s^2LC_2 + s(RLC_2 + CR) + 1} \Rightarrow A(s)$$

$$\circ L(s) = A(s) \cdot B(s) = 1 \Rightarrow L(s) = \frac{-1}{s^2LC_2+1} \circ \frac{g_m RL(s^2LC_2+1)}{s^3LC_1C_2R_L + s^2LC_2 + s(RLC_2 + CR) + 1} \Rightarrow L(s) = \frac{-1}{s^2LC_2+1}$$

$$L(s) = \frac{-g_m R_L}{s^3 L C_1 C_2 R_L + s^2 L C_2 + s R_L (C_1 + C_2) + 1} = 1$$

Critério de Oscilação

$$L(j\omega_0) = \frac{-g_m R_L}{-j\omega_0^3 L C_1 C_2 R_L - L C_2 + j R_L \frac{\omega_0}{C_1 + C_2} (C_1 + C_2) + 1} = 1$$

$$L(j\omega_0) = \frac{-g_m R_L}{j [R_L \omega_0 (C_1 + C_2) - \omega_0^3 L C_1 C_2] - L \omega_0^2 C_2 + 1} = 1$$

$$L(j\omega_0) = \frac{-g_m R_L}{j R_L [\omega_0 (C_1 + C_2) - \omega_0^3 L C_1 C_2] - L \omega_0^2 C_2 + 1} = 1$$

- $L(j\omega_0) = 0 \Rightarrow \omega_0 (C_1 + C_2) - \omega_0^3 L C_1 C_2 = 0$

~~$\omega_0$  Existe~~

$$\omega_0 [C_1 + C_2 - \omega_0^2 L C_1 C_2] = 0$$

$$C_1 + C_2 - \omega_0^2 L C_1 C_2 = 0$$

$$\Rightarrow \omega_0 = \sqrt{\frac{C_1 + C_2}{L C_1 C_2}}$$

- $|L(j\omega_0)| = 1 \Rightarrow \frac{-g_m R_L}{-L \omega_0^2 C_2 + 1} = 1 \Rightarrow g_m R_L = + \sqrt{(C_1 + C_2) C_2}$

$$\Rightarrow g_m R_L = + \frac{(C_1 + C_2) C_2}{C_1} = \boxed{g_m R_L = \frac{C_2}{C_1}}$$

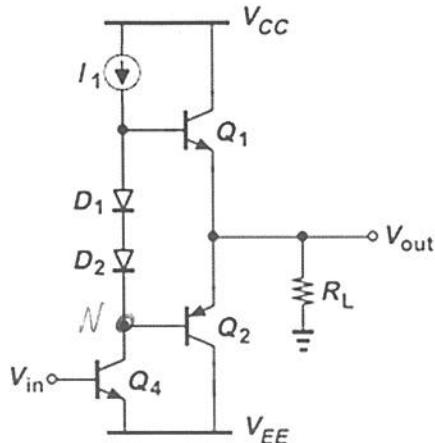
4) a) Para o circuito abaixo , despreze as resistências dos diodos e prove que:

$$A_v = -g_{m4}R_L[(g_{m1} + g_{m2})(r_{\pi 1}/r_{\pi 2}) + 1]$$

b) Para  $g_{m1} \approx g_{m2}$ , prove que (dica: comece reduzindo  $r_{\pi 1}/r_{\pi 2}$  em função de  $g_m$ ):

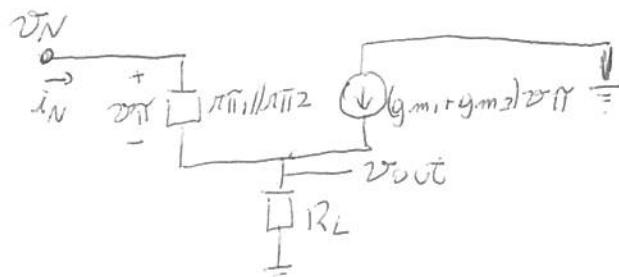
$$A_v \cong -2 \frac{\beta_1 \beta_2}{\beta_1 + \beta_2} g_{m4} R_L$$

c) Dados  $A_v = -4$ ,  $\beta_1 = 40$ ,  $\beta_2 = 20$  e  $R_L = 8\Omega$ , considere  $g_{m1} \approx g_{m2}$  e calcule  $I_{C4}$ .



(a)

Push-Pull ( $A_{v2}$ )



$$v_{out} = R_L \left[ (g_{m1} + g_{m2}) v_{pi} + \frac{v_{pi}}{r_{\pi 1}/r_{\pi 2}} \right]$$

$$v_{out} = v_{pi} R_L \left[ (g_{m1} + g_{m2}) + \frac{1}{r_{\pi 1}/r_{\pi 2}} \right]$$

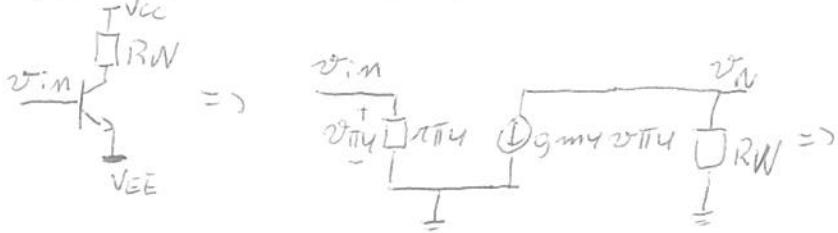
$$v_N = v_{pi} + v_{out} \Rightarrow v_N = v_{pi} \left[ 1 + R_L \left( g_{m1} + g_{m2} + \frac{1}{r_{\pi 1}/r_{\pi 2}} \right) \right]$$

$$A_{v2} = \frac{v_{out}}{v_N} = \frac{R_L \left[ g_{m1} + g_{m2} + \frac{1}{r_{\pi 1}/r_{\pi 2}} \right]}{1 + R_L \left[ g_{m1} + g_{m2} + \frac{1}{r_{\pi 1}/r_{\pi 2}} \right]}$$

$$* R_N \\ i_N = \frac{v_{pi}}{r_{\pi 1}/r_{\pi 2}} \Rightarrow R_N = \frac{v_N}{i_N} = \frac{1 + R_L \left[ g_{m1} + g_{m2} + \frac{1}{r_{\pi 1}/r_{\pi 2}} \right]}{\frac{1}{r_{\pi 1}/r_{\pi 2}}}$$

$$R_N = r_{\pi 1}/r_{\pi 2} \left\{ 1 + R_L \left[ g_{m1} + g_{m2} + \frac{1}{r_{\pi 1}/r_{\pi 2}} \right] \right\}$$

• Emissor comum (Q4, Av1)



$$A_{v1} = \frac{v_{pi}}{v_{in}} = \frac{-g_{m4} v_{pi} R_N}{v_{pi}}$$

$$A_{v1} = -g_{m4} R_N$$

• Ganzes total ( $A\sigma$ )

$$A\sigma = A\sigma_1 \cdot A\sigma_2 = -g_{m4} \cdot \frac{\pi\pi_1/\pi\pi_2}{\pi\pi_1 + \pi\pi_2} \left\{ \frac{1 + RL \left[ g_{m1} + g_{m2} + \frac{1}{\pi\pi_1/\pi\pi_2} \right]}{L + RL \left( g_{m1} + g_{m2} + \frac{1}{\pi\pi_1/\pi\pi_2} \right)} \right\} \cdot \frac{RL \left[ g_{m1}g_{m2} + \frac{1}{\pi\pi_1/\pi\pi_2} \right]}{L + RL \left( g_{m1}g_{m2} + \frac{1}{\pi\pi_1/\pi\pi_2} \right)}$$

$$A\sigma = -g_{m4} RL \left[ (g_{m1} + g_{m2}) \pi\pi_1/\pi\pi_2 + 1 \right] \quad \text{c.q.d.}$$

$$b) \quad \pi\pi_1/\pi\pi_2 = \frac{\pi\pi_1 \pi\pi_2}{\pi\pi_1 + \pi\pi_2} = \frac{\frac{\beta_1}{g_m} \cdot \frac{\beta_2}{g_m}}{\frac{\beta_1}{g_m} + \frac{\beta_2}{g_m}} = \boxed{\frac{1}{g_m} \frac{\beta_1 \beta_2}{\beta_1 + \beta_2}}$$

$$A\sigma = -g_{m4} RL \left[ 2 \cdot g_m \cdot \frac{1}{g_m} \frac{\beta_1 \beta_2}{\beta_1 + \beta_2} \right]$$

$$\underline{2 \cdot \beta_1 \beta_2 \gg 1} \Rightarrow A\sigma \approx -g_{m4} RL \cdot \frac{2 \beta_1 \beta_2}{\beta_1 + \beta_2} \quad \text{c.q.d.}$$

$$c.) \quad -q = -g_{m4} \cdot 8 \cdot 2 \frac{40 \cdot 20}{20 + 40} \Rightarrow g_{m4} = 0,01875$$

$$g_{m4} = \frac{I_{C4}}{V_T \approx 26 \text{ mV}} \Rightarrow I_{C4} = 487,5 \mu\text{A}$$